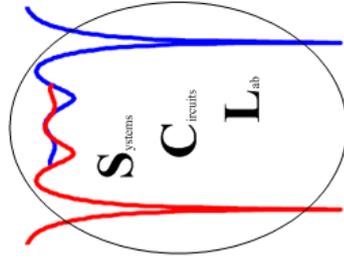


MIMO wireless channel emulation

Roger Piqueras

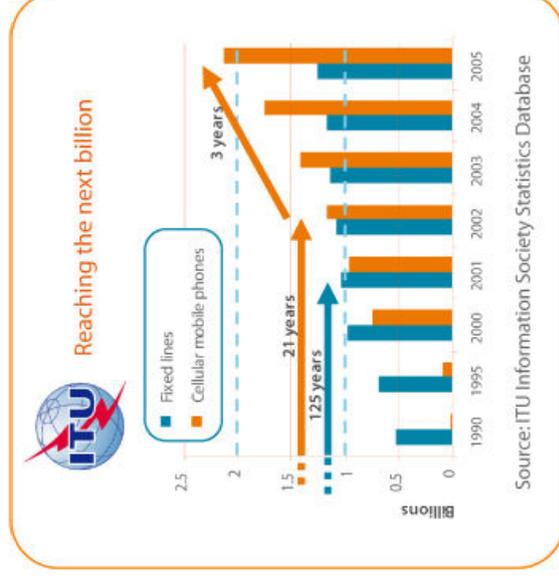
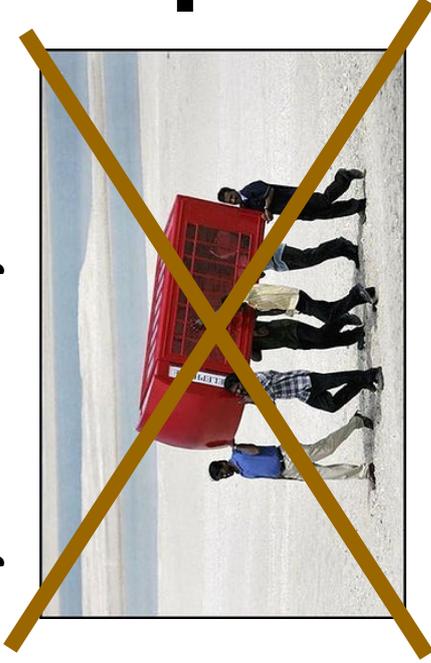


MIMO wireless channel emulation

- Overview:
 - Wireless systems
 - Mobile channel
 - OFDM
 - MIMO
 - IEEE 802.11n
 - Channel models
 - Simulation/Emulation
 - Wireless channel emulation
 - Future work

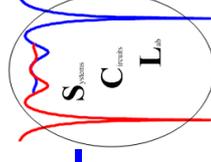
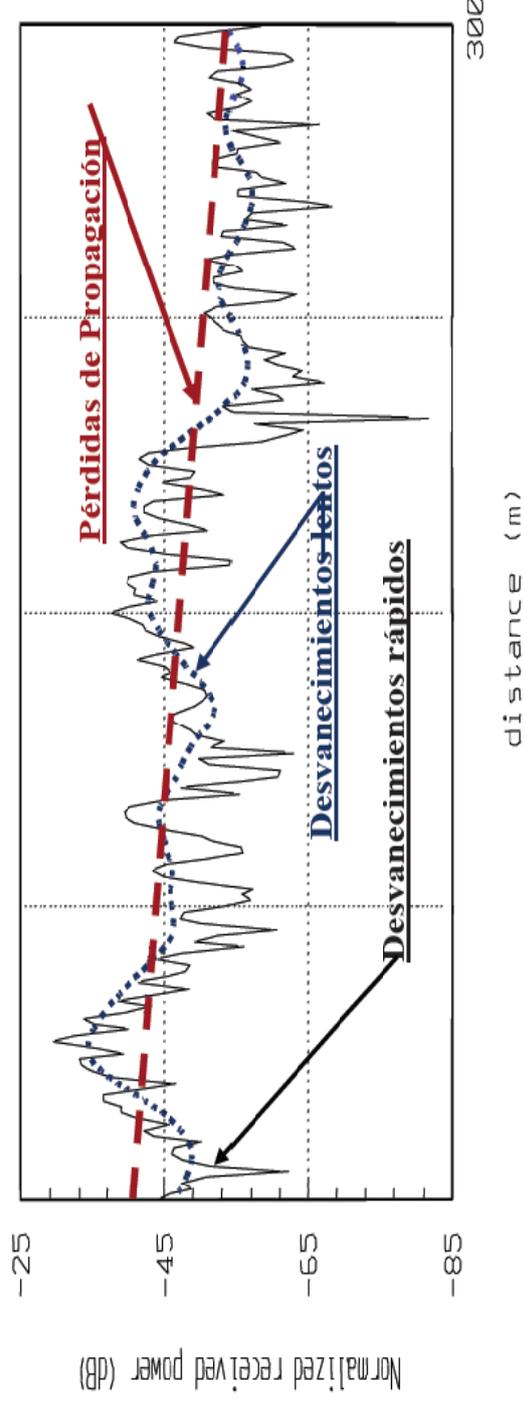
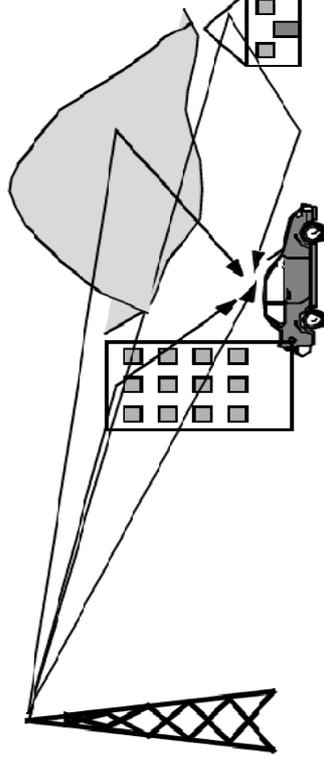
Wireless systems

- Why wireless systems?



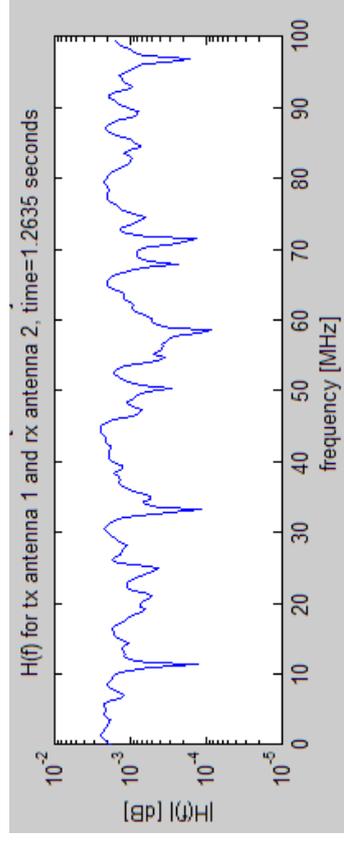
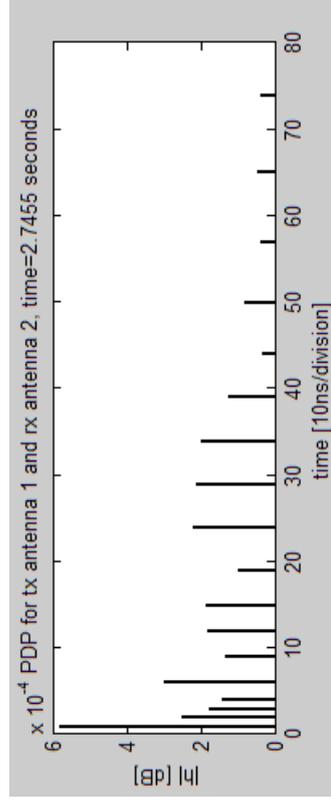
Mobile channel

- The mobile channel is a complex channel:
 - Multipath, scattering
 - Doppler spectrum
 - Shadowing



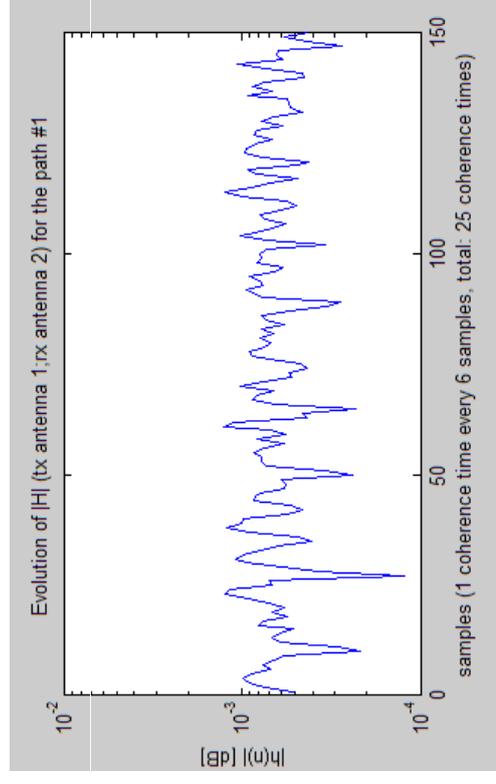
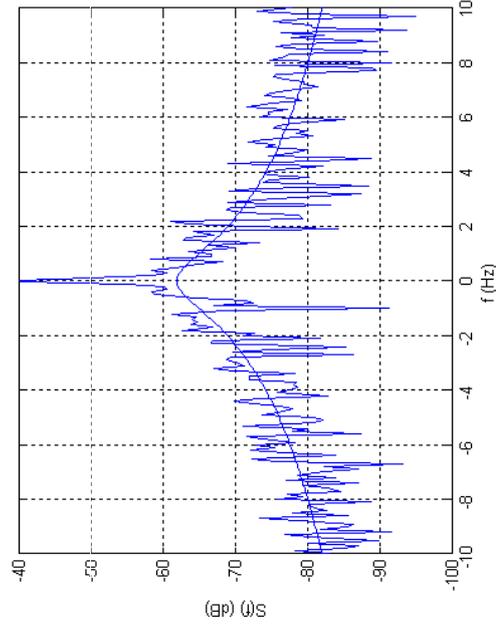
Mobile channel

- Multipath → Frequency selective channel
 - Time domain response of the channel consists on various rays having a complex Gaussian distribution
 - Channel frequency response is not flat
 - Coherence bandwidth



Mobile channel

- Doppler spectrum/fading → Time varying channel
- Change in the carrier frequency due to relative movement between the mobile and the BS
- $f_D \max = v/\lambda$ → Coherence time $\approx 1/f_D \max$
- Fading channel → severe changes in the received signal power every coherence time



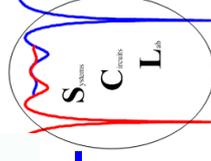
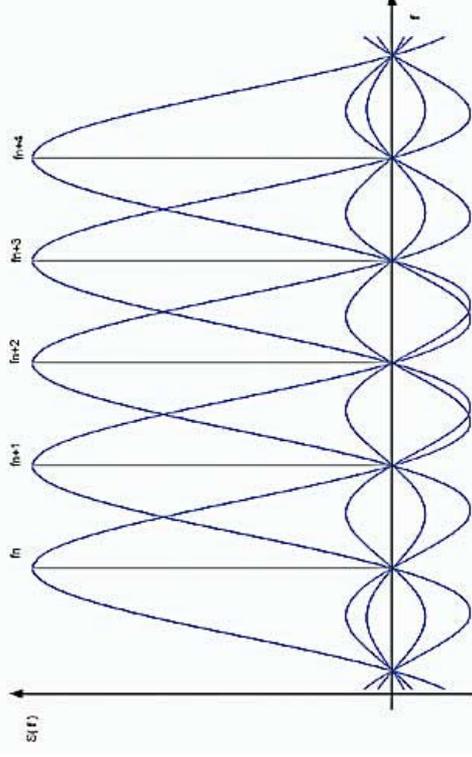
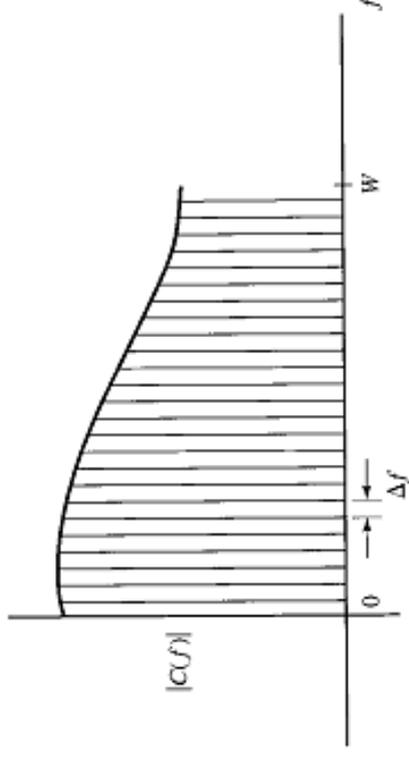
How can we overcome or mitigate the negative effects of the mobile channel?

Mobile channel

- New techniques to handle frequency selectivity and fading channels, such as...
 - OFDM
 - MIMO

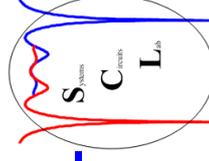
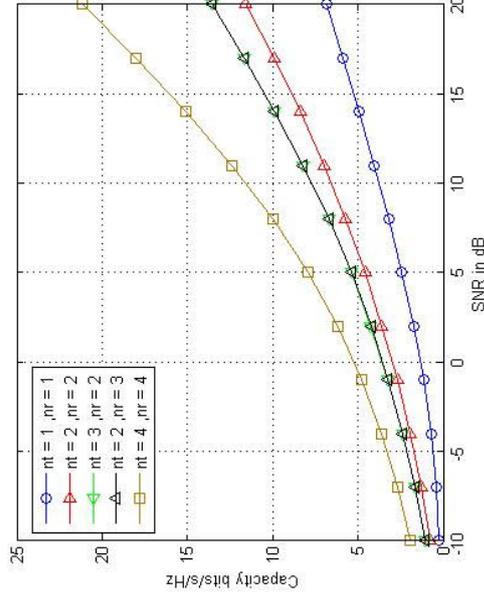
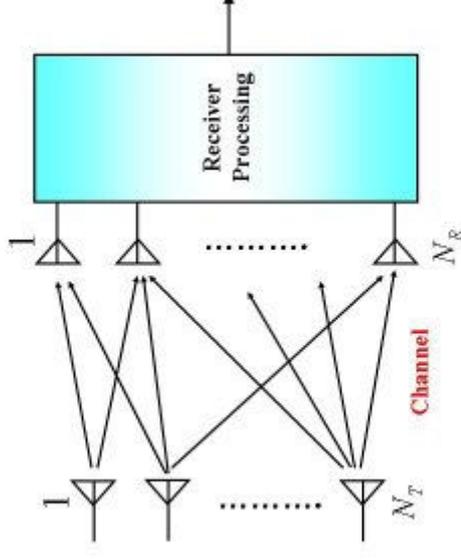
Orthogonal Frequency-Division Multiplexing (OFDM)

- OFDM converts a wideband signal into multiple narrowband signals placed side by side in the frequency domain.
 - Part of the signal is modulated in each subchannel and transmitted through an almost flat fading channel.
 - If one of the subchannels is affected by a severe fading, simply nothing is transmitted in that subchannel
- This results in a strong resiliency against frequency selective channels



Multiple Input-Multiple Output (MIMO)

- Traditionally multiple antennas placed at the receiver to apply diversity
 - Spatial diversity (each antenna will see a different and independent channel)
 - Significant immunity to fading
 - Results improve as the number of antennas increase
- Even better results are achieved when multiple antennas placed also at the transmitter
 - Total available transmit power is split uniformly across transmit antennas
 - Finite optimum number of tx antennas

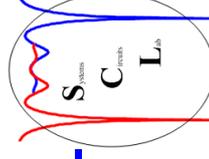
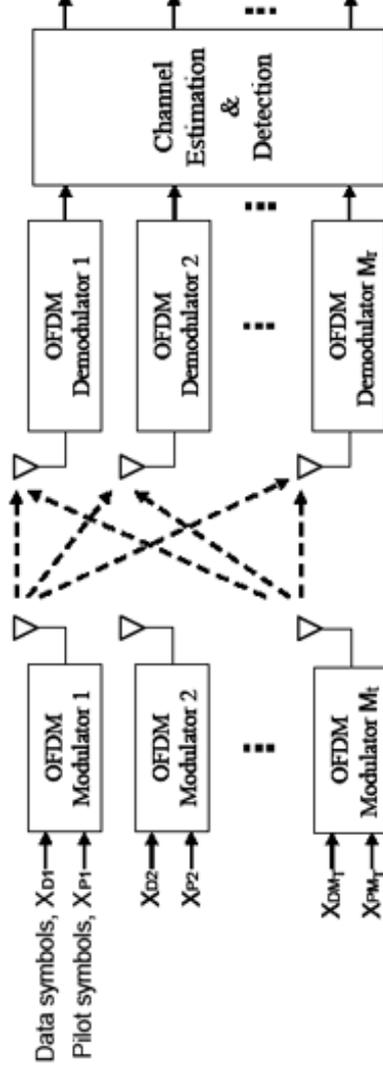


IEEE 802.11n

- OFDM → Resiliency against frequency selective channels
- MIMO → Resiliency against fading channels
- Mobile channel is a fading and frequency selective channel

MIMO+OFDM is a good solution for mobile communications

- IEEE 802.11n standard



IEEE 802.11n

- IEEE 802.11n standard
 - WLAN standard improving the Physical Layer and MAC Layer of the traditional WLAN's
 - Indoor MIMO/OFDM system
 - Range ~ 100 meters
 - Maximum bit rate: 600Mbps (3x3 antennas) (typical bit rate 74 Mbps)
 - Carrier frequency: 2.4 or 5 GHz
 - Full compatibility with current WLANs

Wireless LAN Throughput by IEEE Standard

IEEE WLAN Standard	Over-the-Air (OTA) Estimates	Media Access Control Layer, Service Access Point (SAP) Estimates
802.11b	11 Mbps	5 Mbps
802.11g	54 Mbps	25 Mbps (when .11b is not present)
802.11a	54 Mbps	25 Mbps
802.11n	200+ Mbps	100 Mbps

Table 1. Comparison of different 802.11 transfer rates. (Source: Intel Labs)

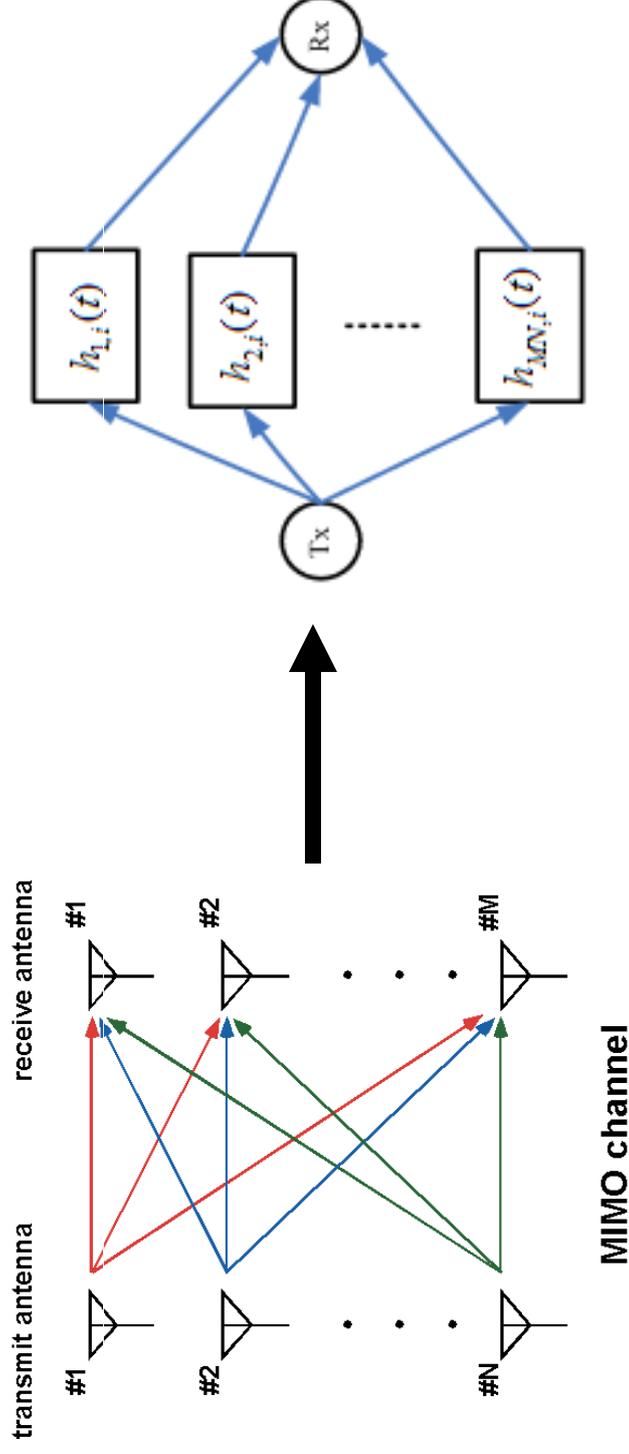
- Project to develop IEEE 802.11n began in 2003
- Just starting to be in a commercial stage
- Intel has started to add 802.11n in some of its products
- For future commercialization, many tests have to be done



Wireless channel on OFDM-MIMO systems simulation

Channel models

- $M \times N$ MIMO channel is equivalent to MN impulse responses, each representing a subchannel
- Model a SISO channel, then extend to MIMO (repeat MN times plus adding other artifacts, such as spatial correlation between antenna elements...)



Channel models

- Statistical Model for Indoor Multipath Propagation (A. Saleh, R. Valenzuela)
- Channel measurement

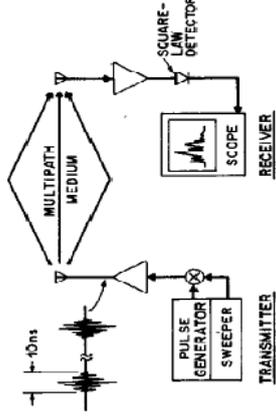
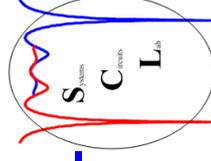
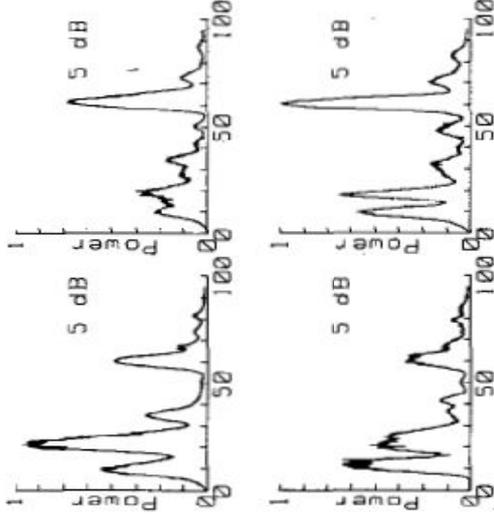


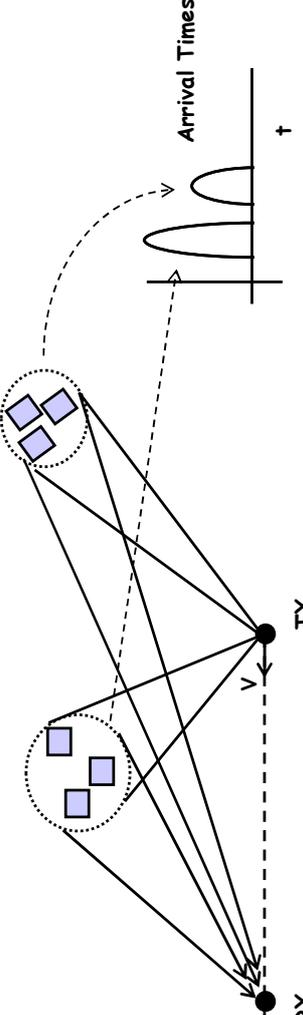
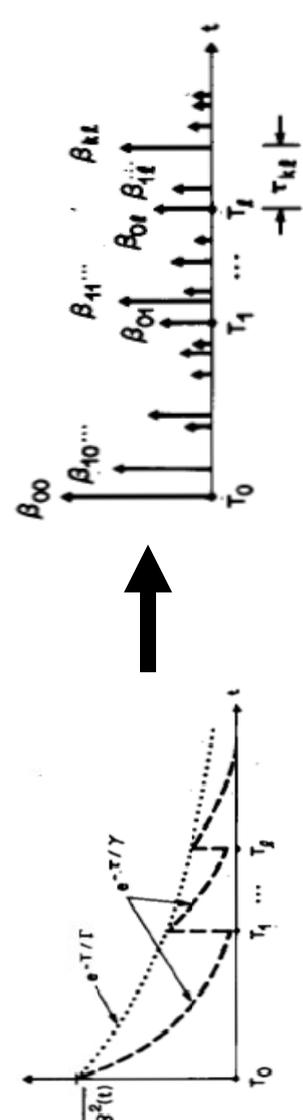
Fig. 2. A schematic representation of the measurements setup.

- Results (Power Delay Profile)

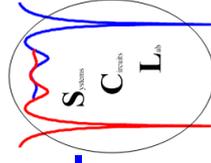
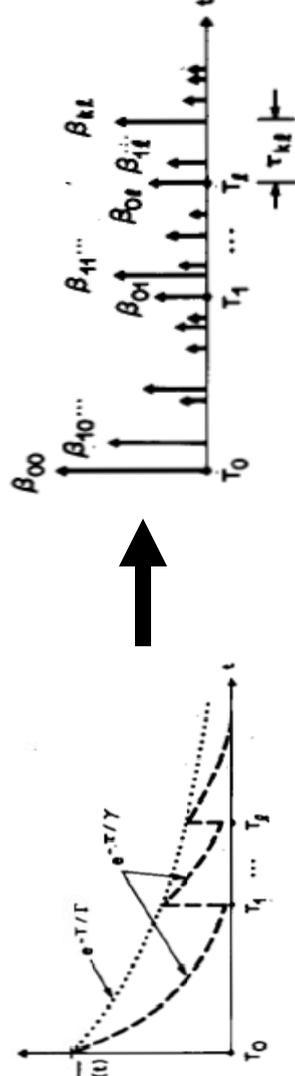


Channel models

- Wireless indoor channel modeling
 - Channel modeled as an FIR

$$h(t) = \sum_{l=0}^{\infty} \sum_{k=0}^{\infty} \beta_{kl} e^{j\theta_{kl}} \delta(t - T_l - \tau_{kl}).$$
 - Multipath: clustering of arriving rays (angular clusters)
 
 - Double exponential decay of the impulses (clusters and rays within a cluster)
 

$$\begin{aligned} \overline{\beta_{kl}^2} &\equiv \overline{\beta^2(T_l, \tau_{kl})} \\ &= \overline{\beta^2(0, 0)} e^{-T_l/\Gamma} e^{-\tau_{kl}/\gamma}, \end{aligned}$$



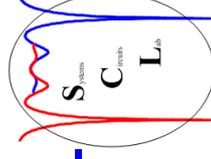
Channel models

- Clustering of rays
 - Doppler spectrum/Fading
 - Shadow fading
 - Path loss model
 - Free space loss (L_{FS}) up to a breakpoint distance (d_{BP})
 - Slope of 3.5 after d_{BP}
- $$L(d) = L_{FS}(d) \quad d \leq d_{BP}$$
- $$L(d) = L_{FS}(d_{BP}) + 35 \log_{10}(d / d_{BP}) \quad d > d_{BP}$$
- Adding these elements together → SISO channel model

Channel models

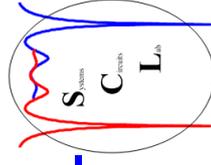
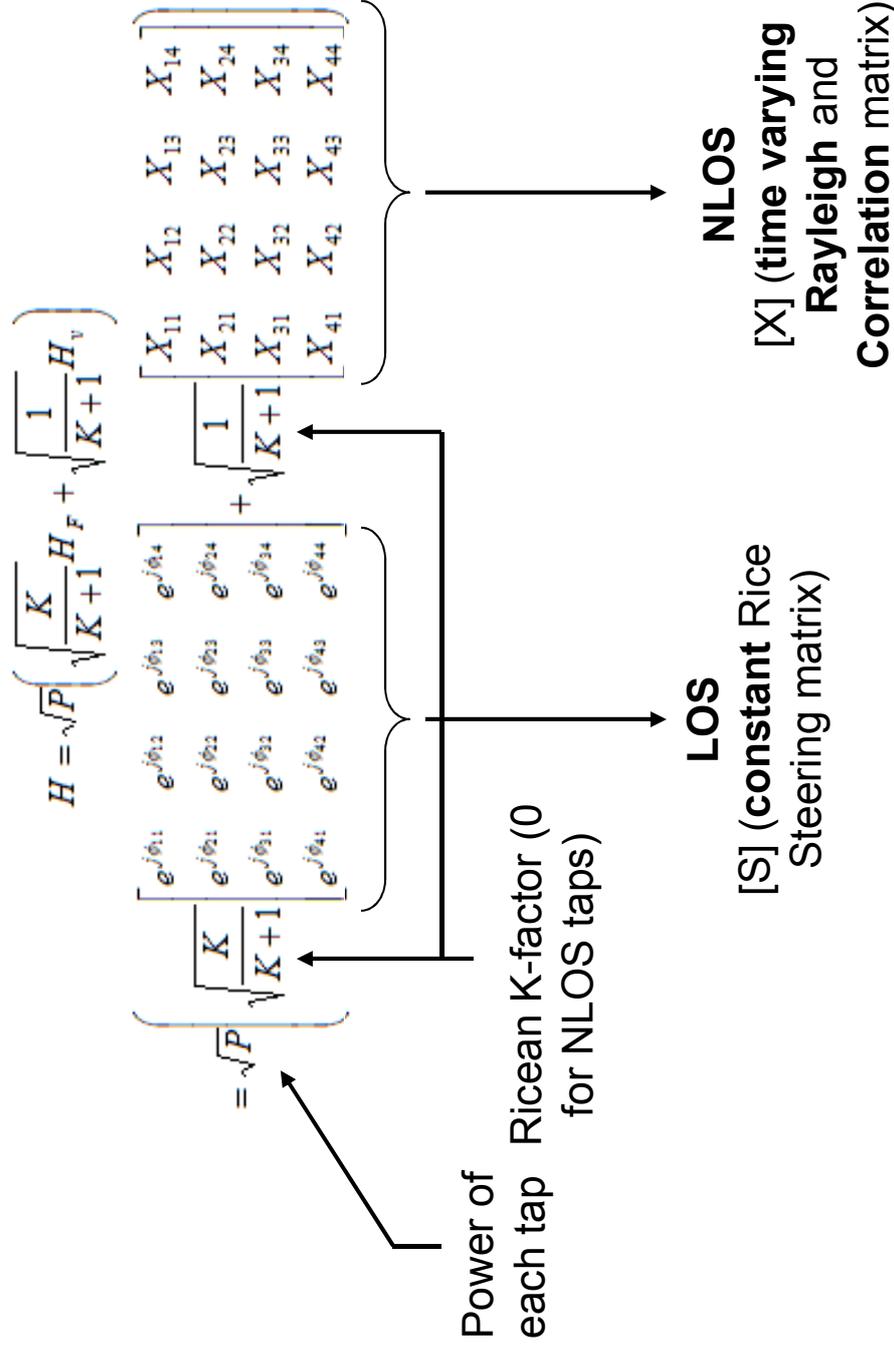
- Set of channel models applicable to indoor WLAN systems
- Channel model is a FIR → Extend to MIMO (matrix of MxN FIRs)
- Other MIMO channel artifacts should be added

Model	Environment	LOS/NLOS	RMS delay spread (ns)	# of clusters
A (optional)	Flat fading	NLOS	0	1 tap
B	Residential	LOS	15	2
C	Residential/ Small Office	LOS/NLOS	30	2
D	Typical Office	NLOS	50	3
E	Large Office	NLOS	100	4
F	Large Space/ (Indoors and Outdoors)	NLOS	150	6



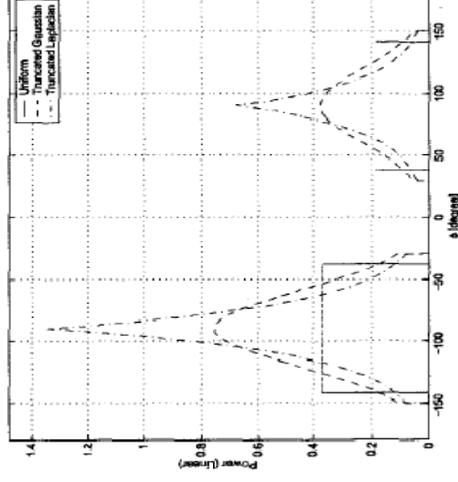
Channel models (MIMO)

- MIMO matrix formulation (for each tap)



Channel models (MIMO)

- NLOS → Rayleigh and Correlation matrix (I)
 - Correlation between tx antennas and rx antennas ([Rtx] and [Rrx] matrixes)
 - From antenna spacings to theoretical capacities (L. Schumacher, K. Pedersen, P. Morgensen)
 - Power Angular Spectrum (PAS): Uniform, Gaussian or Laplacian

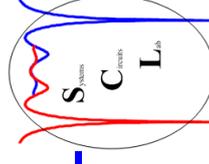


- Uniform linear antenna array → correlation of rx/tx signals between antenna elements

$$\rho_{ij} = \int_{-\pi}^{\pi} e^{j\frac{2\pi}{\lambda}kd \cdot \sin(\theta)} \cdot \underbrace{PAS(\theta) \cdot A_i(\theta) \cdot A_j(\theta)}_{\text{Element electromagnetic pattern}} d\theta$$

Spacing between antennas

Element electromagnetic pattern



Channel models (MIMO)

- NLOS → Rayleigh and Correlation matrix (II)
- Correlation between tx antennas and rx antennas ($[R_{tx}]$ and $[R_{rx}]$ matrixes)
 - Generation of the correlation matrixes

$$[R_{tx}] = [\rho_{txij}]$$

$$[R_{rx}] = [\rho_{rxij}]$$



$$R_{tx} = \begin{bmatrix} 1 & \rho_{tx12}^* & \rho_{tx13}^* & \rho_{tx14}^* \\ \rho_{tx21} & 1 & \rho_{tx23}^* & \rho_{tx24}^* \\ \rho_{tx31} & \rho_{tx32} & 1 & \rho_{tx34}^* \\ \rho_{tx41} & \rho_{tx42} & \rho_{tx43} & 1 \end{bmatrix}$$

$$R_{rx} = \begin{bmatrix} 1 & \rho_{rx12}^* & \rho_{rx13}^* & \rho_{rx14}^* \\ \rho_{rx21} & 1 & \rho_{rx23}^* & \rho_{rx24}^* \\ \rho_{rx31} & \rho_{rx32} & 1 & \rho_{rx34}^* \\ \rho_{rx41} & \rho_{rx42} & \rho_{rx43} & 1 \end{bmatrix}$$

Example: 4x4 MIMO channel

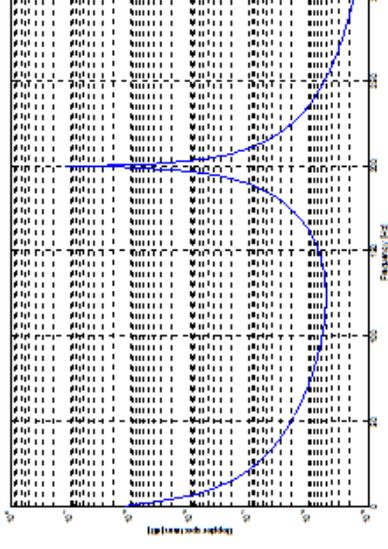
Channel models (MIMO)

- NLOS → Rayleigh and Correlation matrix (III)
 - Rayleigh fading
 - Generation of a Rayleigh vector
 - Shaping of the vector with a FIR (order=7) to generate the adequate Doppler spectrum
 - Combination of the Correlation and the Rayleigh fading

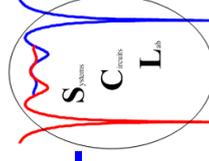
$$[X] = \{ [R_{tx}] \otimes [R_{rx}] \}^{1/2} [H_{iid}]$$

Kronecker product → Rayleigh vector

- Adding other artifacts
 - Shadow fading
 - Doppler component due to a moving vehicle (optional)
 - Doppler components due to fluorescent lights (optional)



Bell shape doppler spectrum with a 200Hz Doppler component



Simulation/Emulation

- IEEE 802.11n still in testing stage (already being implemented by Intel)
- Performance of advanced wireless techniques such as MIMO is highly dependant on the channel
 - Necessity of realising credible tests and validations under realistic conditions
 - Inefficiency of real measurements (moving the rx along a specific indoor location)

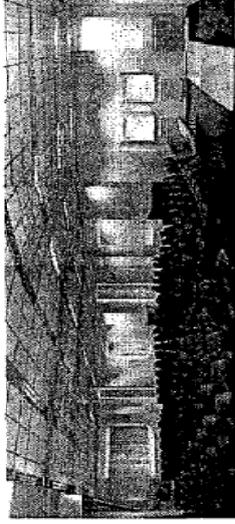


Figure 4: Location used for Empty Room measurements

D.P. McNamara, M.A. Beach, P.N.
Fletcher, P. Karlsson, Initial
Investigation of MIMO Channels in
Indoor Enviroments

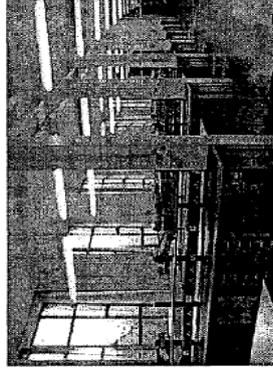


Figure 2: Laboratory measurement environment

Necessity of replicating the wireless channel in the laboratory

Simulation/Emulation

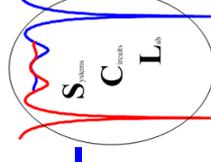
- Wireless channel replication in the laboratory
 - Traditional option → Simulation
 - “Easy” solution, but...

$$H = \sqrt{P} \left(\sqrt{\frac{K}{K+1}} H_F + \sqrt{\frac{1}{K+1}} H_V \right)$$

$$= \sqrt{P} \begin{bmatrix} \sqrt{\frac{K}{K+1}} e^{j\phi_{11}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{12}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{13}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{14}} \\ \sqrt{\frac{K}{K+1}} e^{j\phi_{21}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{22}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{23}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{24}} \\ \sqrt{\frac{K}{K+1}} e^{j\phi_{31}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{32}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{33}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{34}} \\ \sqrt{\frac{K}{K+1}} e^{j\phi_{41}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{42}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{43}} & \sqrt{\frac{K}{K+1}} e^{j\phi_{44}} \end{bmatrix} + \sqrt{\frac{1}{K+1}} \begin{bmatrix} X_{11} & X_{12} & X_{13} & X_{14} \\ X_{21} & X_{22} & X_{23} & X_{24} \\ X_{31} & X_{32} & X_{33} & X_{34} \\ X_{41} & X_{42} & X_{43} & X_{44} \end{bmatrix}$$

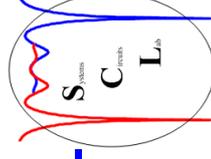
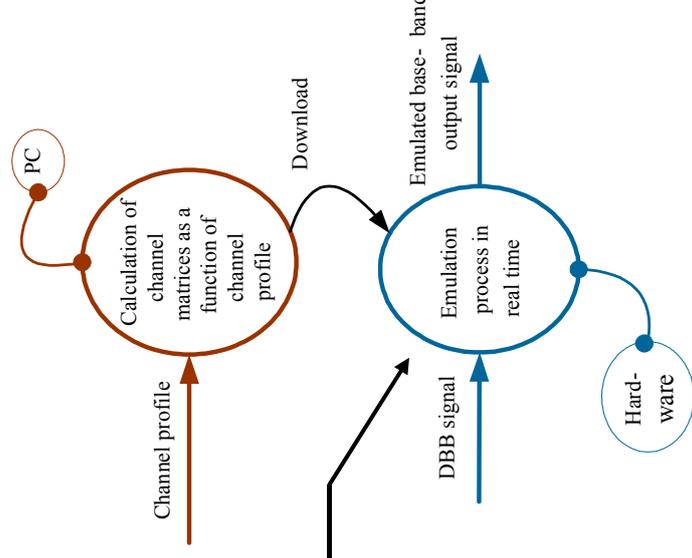
- MIMO systems → **The channel matrix [H] has to be generated at least every coherence time!**
- A more realistic simulation will require many samples per coherence time
- **Huge computational complexity** (that even grows quadratically with the number of tx and rx antennas)

Simulation is not enough → Real time emulation



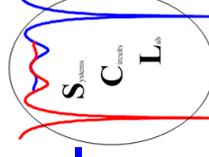
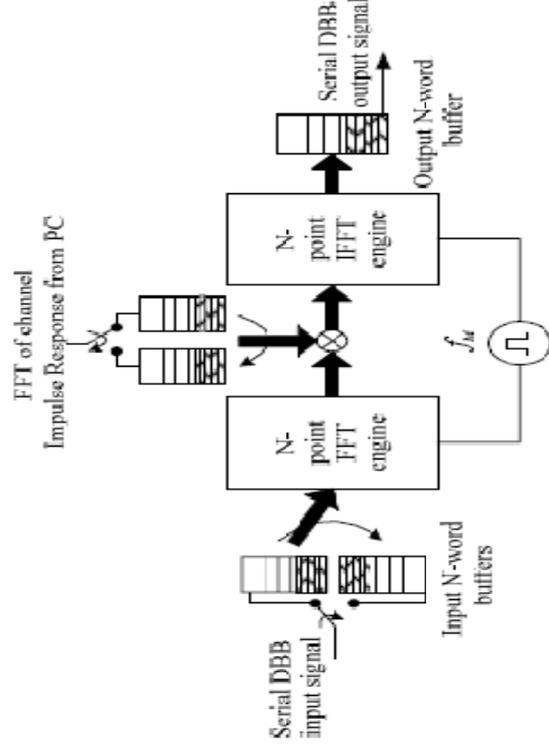
Wireless channel emulation

- Channel emulator
 - A platform that replicates an actual wireless channel and all its artifacts in the lab environment
 - Due to the large amount of real-time operations demanded, usually implemented as high speed dedicated hardware platforms
- Traditional approach
 - Represent the channel as a FIR filter compliant to the PDP
 - Emulation core performs convolution in time over the incoming data and the channel response
 - Replicate for the number of subchannels (MxN)
 - Computations are a function of the channel response length and increase quadratically with M and N



Wireless channel emulation

- New scalable solutions → Frequency domain channel emulation
- A real-time wireless channel emulator for MIMO systems (H. Eslami, A. Eltawil)
 - Generate the frequency response for each subchannel ($H_{ij}(f)$)
 - One sample of the channel state every coherence time
 - For accurate emulations, multiple samples can be generated every coherence time
 - 1 FFT/IFFT per each tx/rx element (M FFT, N IFFT) as oppose to one FIR per each subchannel
- Emulator core is a single multiplier for each subchannel

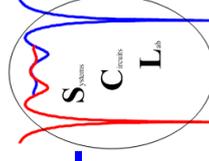
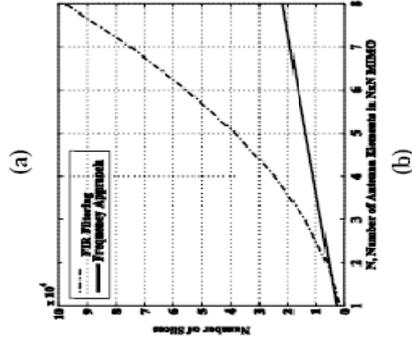
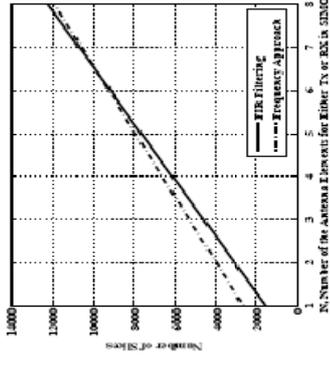
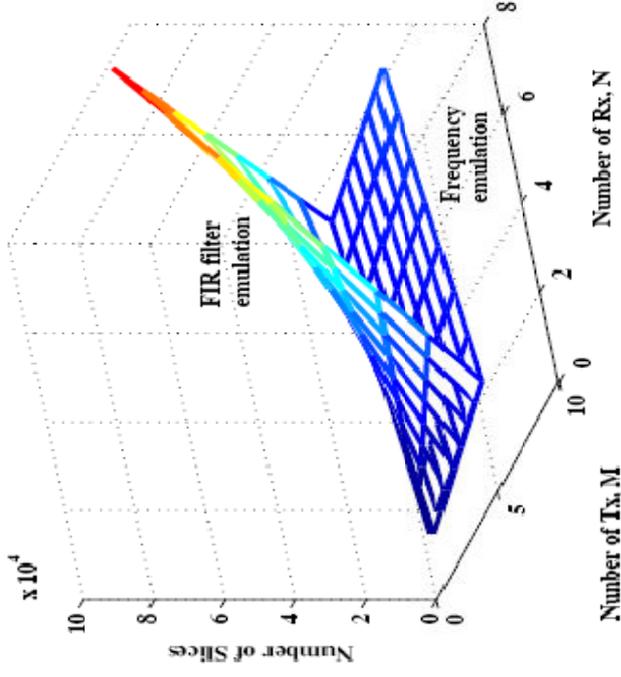


Wireless channel emulation

- Improvements
 - Each extra sub-channel is only an additional multiplier
 - Less computational complexity
 - Scalable (no changes needed if the length of the channel FIR varies)
- For example, for a 4x4 MIMO system
 - 4 FFT engines at the input of the emulator
 - 16 multipliers (1 per each subchannel)
 - 4 IFFT engines at the output of the emulator

Wireless channel emulation

- Some results...
- FPGA implementation of the time domain emulator vs the frequency domain emulator



Future work

- Optimize and reduce computational complexity
 - Hybrid time/frequency channel emulator
- ...